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(54) Improvement to planar antennas of the slot type

(57) The present invention relates to a planar antenna carried by a substrate including a slot (11) consisting of a closed curve dimensioned in order to operate at a given frequency and fed by a feed line (12) positioned so that the slot lies in a short-circuit plane of the feed

line. This antenna includes, in parallel on the slot 11, at least one switching means 13 capable of assuming a closed state or an open state so as to modify the operating frequency band of the planar antenna. This antenna is particularly suitable for domestic wireless networks.

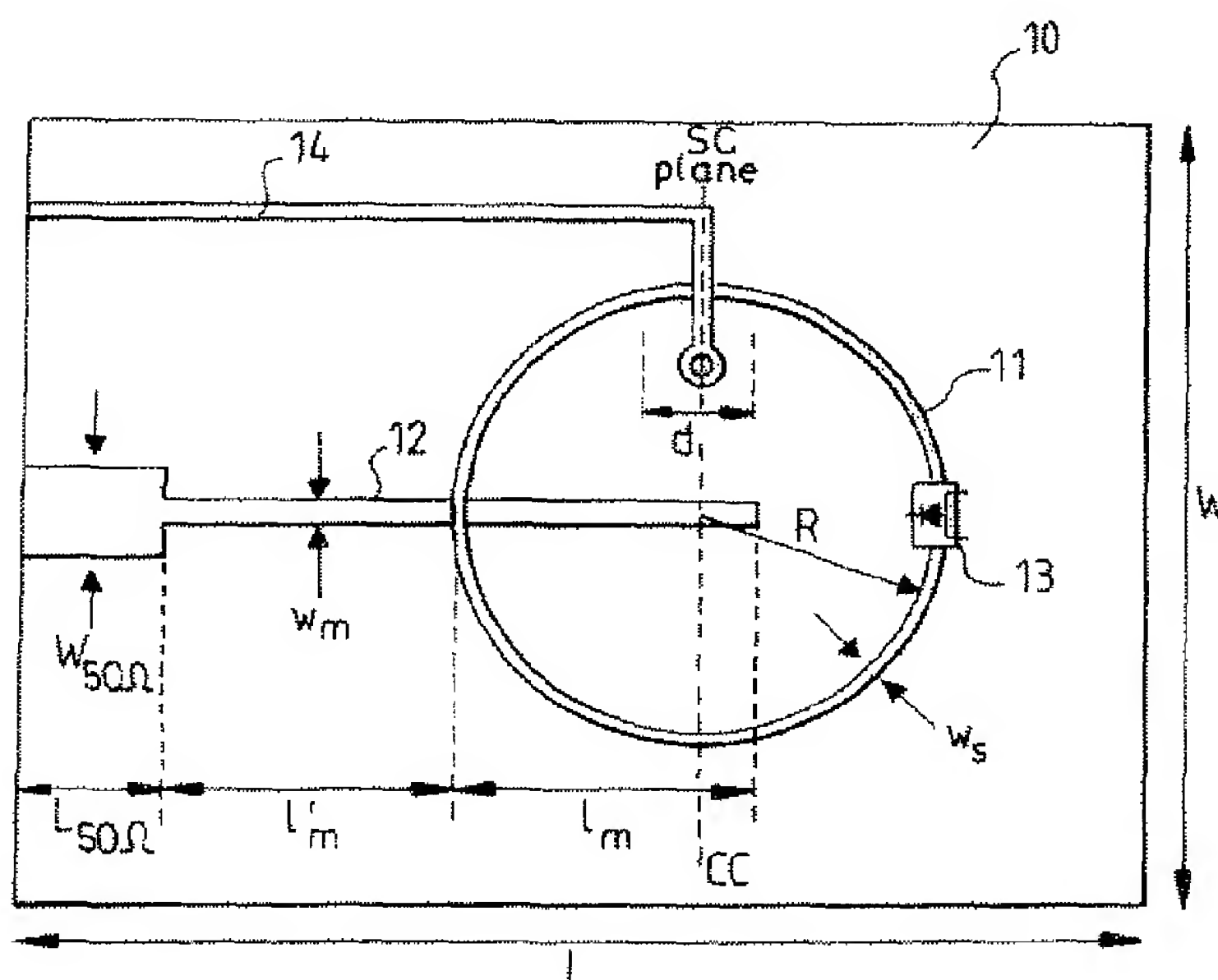


FIG.3

Description

[0001] The present invention relates to a planar antenna, more particularly to a multiband planar antenna of the slot type suitable for wireless networks, in particular for wireless networks operating in separate frequency bands.

[0002] In the scope of deploying mobile or domestic wireless networks, the design of the antennas is confronted with a particular problem which stems from the way in which the various frequencies are allocated to these networks. For instance, in the case of domestic wireless networks in the IEEE802.11a or Hiperlan2 standard, two separate frequency blocks operating in the 5 GHz band have been allocated to the various operators, as can be seen from the table below.

Table A

Technology	Application	Frequency band (GHz)
Europe BRAN/ HYPERLAN2	Domestic networks	(5.15-5.35) (5.47-5.725)
US-IEEE 802.11a	Domestic networks	(5.15-5.35) (5.725-5.825)

[0003] In order to cover both frequency bands, whether for a single standard or for both standards simultaneously, a variety of solutions have been proposed. The most obvious solution consists in using an antenna with a wide frequency band which covers both frequency bands at the same time. This type of wide-frequency-band antenna is generally complex in structure and high in cost. The use of a wide-band antenna also has other drawbacks, such as the degradation of the performance of the receiver due to the noise bandwidth and the jammer which can operate throughout the band covered by the antenna, this band also including the band unallocated to the specific applications which lie between 5.35 GHz and 5.47 GHz. Using a wide-frequency-band antenna involves more stringent filtering constraints for the transmitter, in order to comply with the out-of-band transmission power masks or profiles, namely the maximum powers which are allowed to be transmitted inside the allocated band, but also outside this band. This leads to additional losses and extra cost for the equipment.

[0004] In wireless networks, at a given instant, the antenna furthermore covers a channel having a width of about 20 MHz, lying in one or the other of the two bands. One solution making it possible to avoid the drawbacks associated with wide-frequency-band antennas might be to use an antenna whose frequency band can be tuned electronically.

[0005] Planar antennas which, as represented in Figure 1, consist of an annular slot 1 operating at a given frequency f are also known, the slot being fed by a feed line 2. More precisely, on a substrate consisting of a usual printed circuit metallized on both of its faces, the annular slot 1 which may be circular in shape, but which may also have any other closed shape, is produced conventionally by etching on the side intended to constitute the earth plane of the antenna. The feed line 2 is intended to feed the slot 1 with energy by electromagnetic coupling. For example, it consists of a line produced in microstrip technology, which is positioned on the other side of the substrate from the slot 1 and is oriented radially with respect to the circle which forms this slot, in the embodiment which is represented.

[0006] In this embodiment, the microstrip line-annular slot transition of the antenna is produced in a known fashion so that the slot 1 lies in a line short-circuit plane, that is to say in a region where the currents are strongest. Hence, $l_m = k\lambda_m/4$, where λ_m is the wavelength being guided in the line and k is an odd integer. The length l_m is chosen in order to achieve 50Ω matching of the line 2. In this case, the perimeter p of the slot 1 is chosen to be equal to a multiple m of the wavelength being guided in the slot, m being a positive whole number. Hence, $P = 2\pi R = m\lambda$, where λ is the wave length being guided in the slot. In this case, the resonant frequencies of the various modes are in practice multiples of the frequency f , these modes corresponding to the fundamental mode, the higher mode etc.

[0007] An antenna of this type can hence be modelled around its resonant frequency f by a parallel RLC circuit, such as represented in Figure 2. The relationship $LC\omega^2 = 1$ is therefore obtained at the resonant frequency, with $\omega = 2\pi f$, f being equal to the resonant frequency.

[0008] The antenna described above offers the particular advantage of having a compact structure and of being easy to produce. It is furthermore known to the person skilled in the art that the equivalent circuit of a diode, in particular a PIN diode, is a capacitive circuit when the diode is in the OFF state or an inductive circuit when the diode is in the ON state.

[0009] The present invention therefore relates to an improvement to planar antennas of the annular slot type, which makes it possible to provide coverage of a plurality of frequency bands while avoiding the drawbacks and difficulties associated with wide-frequency-band antennas.

[0010] The present invention hence relates to a planar antenna carried by a substrate including a slot consisting of a closed curve dimensioned in order to operate at a given frequency and fed by a feed line positioned so that the slot lies in a short-circuit plane of the feed line, characterized in that it includes, in parallel on the slot, a plurality of switching means capable of assuming a closed state or an open state so as to modify the central frequency and the width of the operating frequency band of the planar antenna.

[0011] The switching means preferably consist of a diode or a varactor allowing continuous adjustment of the frequency. According to an alternative embodiment, a diode is at least put in parallel with a varactor. Furthermore, the switching means is or are fitted in parallel, as a function of the resonant frequency desired for the antenna, between the electrical short-circuit plane for the slot, giving a minimum value, and the electrical open-circuit plane for the slot, giving a maximum value.

[0012] Other characteristics and advantages of the present invention will become apparent on reading the description given below of a preferred embodiment with reference to the drawings in which:

Figure 1, already described, represents a planar antenna of the annular slot type according to the prior art.

Figure 2 is an equivalent circuit diagram of the antenna in Figure 1.

Figure 3 is a plan view of one embodiment of the present invention.

Figures 4a and 4b are equivalent circuit diagrams of the antenna in Figure 3.

Figure 5 represents the reflection coefficient as a function of frequency of the antenna in Figure 3, when the diode is in an open-circuit plane for the slot, for both states of the diode: ON or OFF.

Figure 6 is a schematic plan view of an antenna according to the present invention, showing various possible positions for the diode.

Figure 7 represents a curve giving the reflection coefficient as a function of frequency for the various possible positions for the diode.

Figure 8 is a schematic plan view of an annular slot-type antenna provided with two diodes on either side of the short-circuit plane, according to another embodiment of the present invention.

Figure 9 is a diagram giving the reflection coefficient as a function of frequency for the antenna in Figure 8 for both states of the diode.

[0013] To simplify the description in the figures, the same elements bear the same references.

[0014] A first embodiment of the present invention will be described first with reference to Figures 3 to 5. Hence, as represented in Figure 3, the planar antenna according to the present invention consists of an annular slot 11 produced in a known fashion on a substrate 10. This annular slot 11 is fed by a feed line 12, more particularly a microstrip line connected to a radiofrequency feed. Furthermore, as represented in Figure 3, a feed line 14 terminated by a metallized hole provides the continuous control of the antenna. An antenna of this type was produced for the measurements. In this case, the antenna is produced on an R04003 substrate having a height $h = 0.81$ mm, a dielectric constant $\epsilon_r = 3.38$ and a tangent $\delta = 0.0022$. In this case, the substrate which is metallized in a known fashion forms an earth plane of length $L = 35$ mm and of width $W = 30$ mm. The annular slot has a radius $R = 6.7$ mm, a width $W_s = 0.4$ mm. The microstrip line 12 is placed so that the slot 11 lies in a short-circuit plane of the feed line. Therefore, the feed line 12 overlaps the slot 11 by a length $l_m = k\lambda_m/4$, where λ_m is the wavelength being guided in the line and k is an odd integer. In the present case, $l_m = 8.5$ mm. The width of the line 12 $W_m = 0.3$ mm. Furthermore, the feed line 12 is terminated by a length of 50Ω impedance line matched to the standard impedance of a connector, such that $L_{50\Omega} = 4.8$ mm and $W_{50\Omega} = 1.85$ mm.

[0015] According to the present invention, a diode 13, namely a PIN diode such as the HP diodes Ref: HSMP-489B in the embodiment which is represented, is positioned in parallel on the slot 11. In the embodiment of Figure 3, the diode 13 is placed in an open-circuit plane of the slot 11. This diode 13 is connected to a control circuit (not shown) for allowing it to be put either into an OFF state or into an ON state.

[0016] The operation of an antenna of the type having an annular slot, provided with a diode in parallel, will now be explained more particularly with reference to Figures 4a and 4b.

[0017] Knowing that when a diode is in the OFF state, its operation is capacitive operation, a circuit equivalent to that in Figure 4a is therefore obtained in this case, namely two capacitors C and C_d in parallel giving a capacitance C_e whose value is such that $C_e = C + C_d$. In the known fashion, the resonant frequency f' of this circuit is given by the condition $LC_e\omega'^2 = 1$, with $\omega' = 2\pi f'$. Since C_e has a value higher than the value C corresponding to the slot without any diode, it can be deduced therefrom that the frequency f' is lower than the frequency f of the slot without any diode.

[0018] Knowing that a diode in the ON state has inductive operation, a diagram equivalent to that in Figure 4b is obtained, in which the two inductances L and L_d are in parallel. In this case, the value L_e of the equivalent inductance is equal to $L_e = LL_d/(L+L_d)$. In this circuit, the operating frequency f'' is given by the new resonance condition $L_e C\omega''^2 = 1$, with $\omega'' = 2\pi f''$. Since L_e is less than L , it can be deduced that the frequency f'' is higher than the frequency f of the slot without any diode. By controlling the state of the diode 13, it is hence possible to control the resonant frequency of the antenna in Figure 3.

[0019] The effect of putting a plurality of diodes in parallel will therefore be:

1/ to increase the difference between the low frequency f' obtained for diodes in the OFF state and the frequency f in the absence of any diode,

2/ to increase the difference between the high frequency f'' obtained for diodes in the ON state and the frequency f in the absence of any diode.

[0020] It is therefore possible to control the resonant frequency of the antenna in Figure 3 over bands which are more or less wide and are more or less symmetrical with respect to the resonant frequency of a slot in the absence of any diode.

[0021] The curve in Figure 5 clearly shows, for the antenna structure in Figure 3, that switching the PIN diode 13 from an OFF state to an ON state makes it possible to change from a frequency of about 4.8 GHz, for the diode in the OFF state, to a frequency of about 7.1 GHz for a diode in the ON state.

[0022] The effect produced by the placement of the diode or diodes in the slot will now be shown with reference to Figures 6 and 7, this effect leading to an influence on the operating frequency of the slot.

[0023] Hence, Figure 6 schematically represents an annular slot 11 fed, for example, by a microstrip line 12. In this figure, the diode is fitted in parallel in the slot at various positions between a position corresponding to an open-circuit plane, as for the diode 13, and a position corresponding to a short-circuit plane, as for the diode 13'. The other diodes are positioned, for example, at 22°, 45° and 60° from the short-circuit plane. The coupling of the diode with the resonant slot 11 is modified in this case, which modifies the exact value of the equivalent capacitance, in the case of an OFF state, or of the inductance in the case of ON state. When the diode 13' is placed in an electrical short-circuit plane, it hence contributes an impedance (inductive or capacitive, depending on the state) in parallel with a zero impedance. Its effect is therefore minimal. When the diode 13 is placed in an open-circuit plane, conversely, it contributes an impedance parallel with infinite impedance and its effect is maximum. The various results obtained are represented in Figure 7, which gives the reflection coefficient S11 in dB as a function of the frequency in GHz.

[0024] Figures 8 and 9 represent an alternative embodiment of the present invention. Figure 8 represents a planar antenna consisting, as Figure 3, of a slot antenna 11 fed by a microstrip line 12, a microstrip line 14 controlling the continuous value of the antenna. In this case, as represented in Figure 8, two diodes 15A, 15B are fitted in parallel on the slot on either side of the short-circuit plane for the slot, referenced SC plane. In this embodiment, the distance d between the two diodes 15A, 15B is equal to 2.8 mm. When the diodes change from the OFF state to the ON state in this case, the operating frequency changes from 5.54 GHz to 5.94 GHz as represented in Figure 9, which gives the reflection coefficient S11 in dB as a function of the frequency in GHz. A frequency shift of 500 MHz is therefore observed.

[0025] Radiation diagram measurements were furthermore carried out in an anechoic chamber with an antenna model as represented in Figure 8, and having the dimensions given above. It is found in this case that the diodes do not perturb the basic radiation of the annular slot.

[0026] The present invention has been described with reference to PIN diodes as the switching means. It is clear to the person skilled in the art that other switching means may be used, in particular varactors which allow quasi-continuous control of the resonant frequency in a given frequency range. Specifically, a varactor is an electronic component (typically a reverse-biased diode) which makes it possible to control the junction capacitance (OFF-state diode) which decreases as a function of the voltage applied to its terminals. It is hence possible to modify the resonant frequency of the antenna continuously by modifying the bias voltage of the varactor. The varactors may be associated with at least one of the PIN diodes, so as to allow quasi-continuous frequency control over one or more ranges. The slot may furthermore have a closed shape other than an annular shape. It may have a polygonal shape such as square, triangular, rectangular. The invention described above therefore provides a compact and inexpensive planar antenna which can operate in multiple frequency bands corresponding, in particular, to the IEEE802.11a or Hyperlan2 standard.

Claims

1. Planar antenna carried by a substrate including a slot (11) consisting of a closed curve dimensioned in order to operate at a given frequency and fed by a feed line (12) positioned so that the slot lies in a short-circuit plane of the feed line, **characterized in that** it includes, in parallel on the slot, a plurality of switching means (13; 13, 13'; 15A, 15B) capable of assuming a closed state or an open state so as to modify the central frequency and the width of the operating frequency band of the planar antenna.
2. Antenna according to Claim 1, **characterized in that** the switching means are fitted, in parallel as a function of the desired resonant frequency of the antenna, between the electrical short-circuit plane for the slot, giving a minimum value, and the electrical open-circuit plane for the slot, giving a maximum value.
3. Antenna according to either one of Claims 1 and 2, **characterized in that** the switching means consists of a diode.
4. Antenna according to either one of Claims 1 and 2, **characterized in that** switching means consists of a varactor.

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5. Antenna according to either one of Claims 3 and 4, **characterized in that** at least one diode is put in parallel with a varactor.
6. Antenna according to any one of the preceding claims, **characterized in that** the slot is annular or polygonal in shape.

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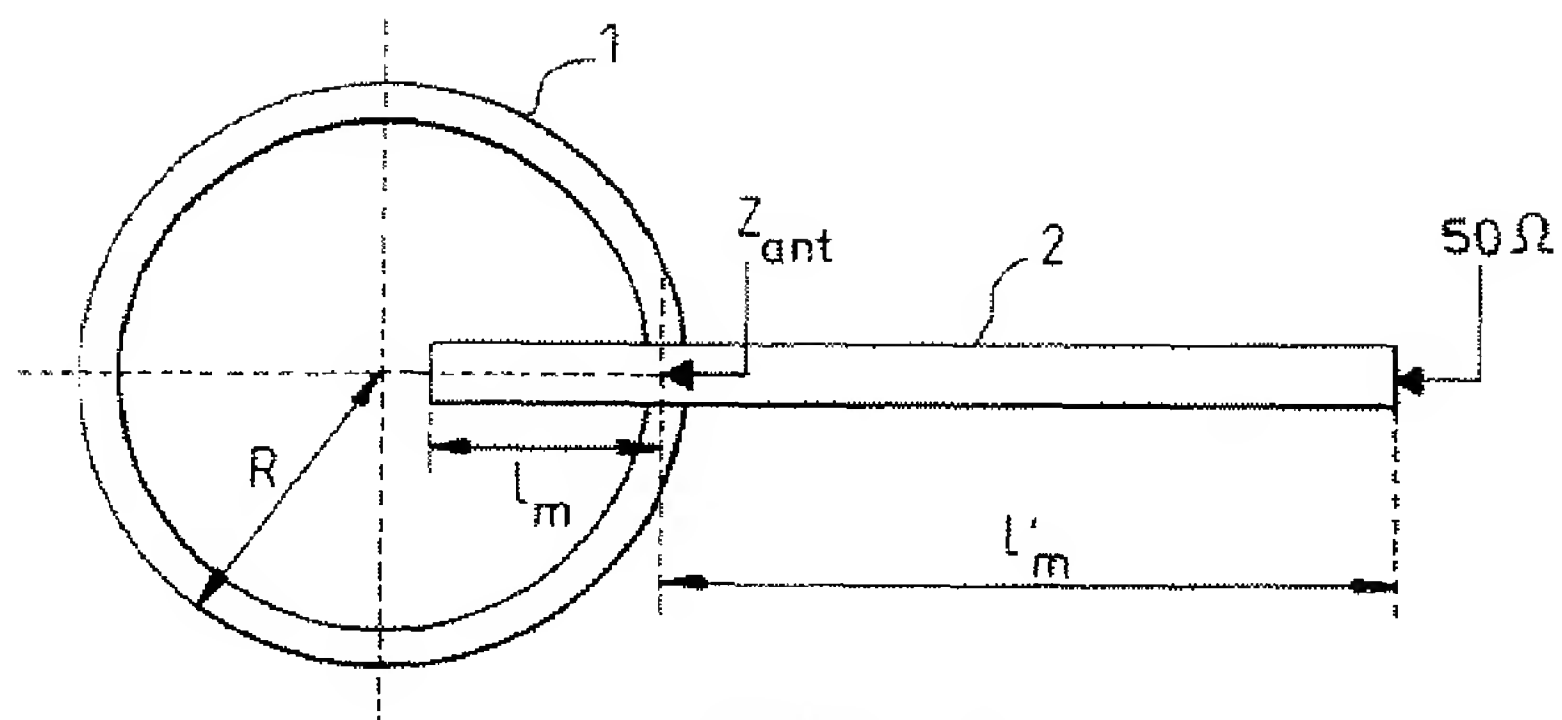


FIG. 1

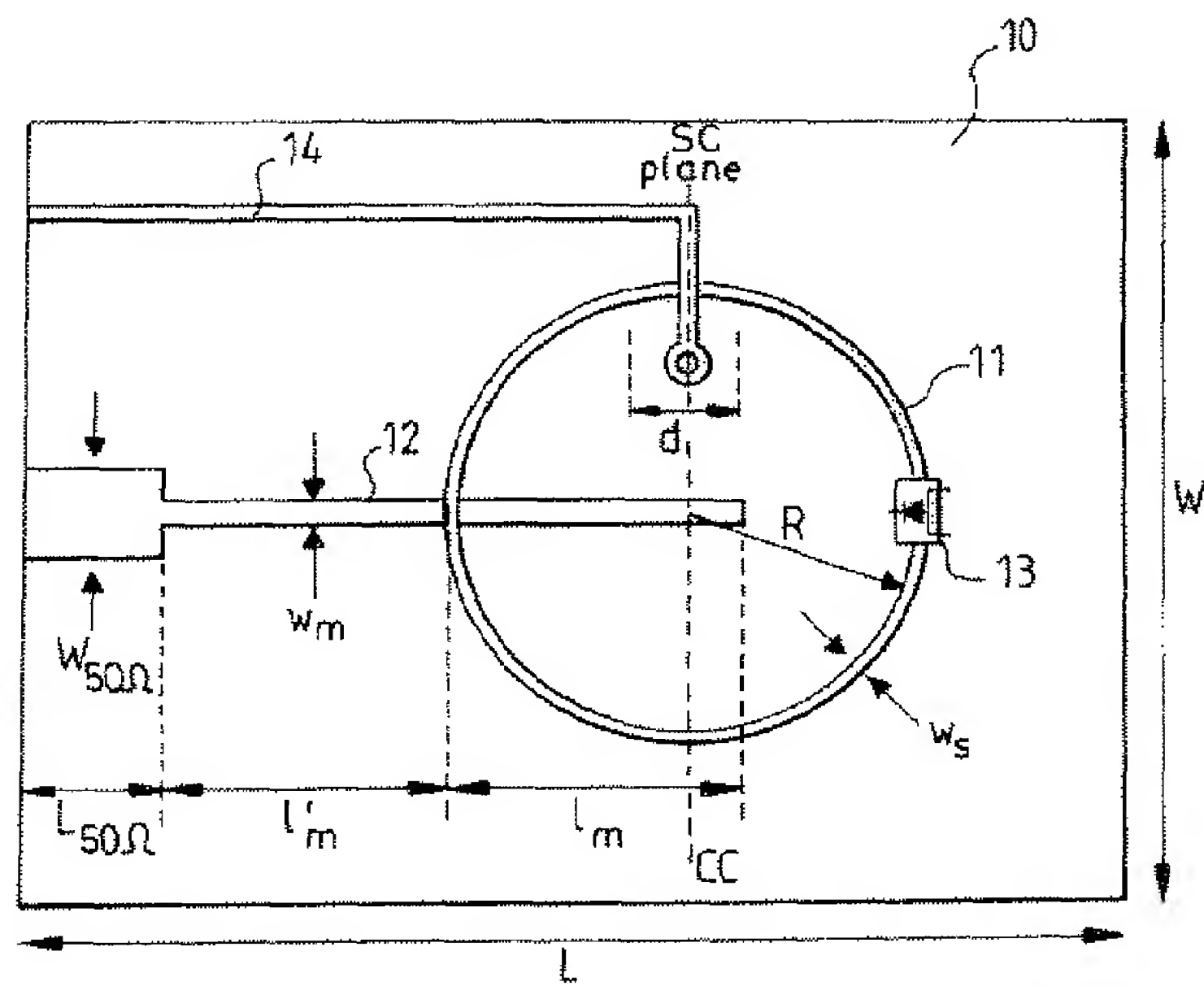


FIG. 3

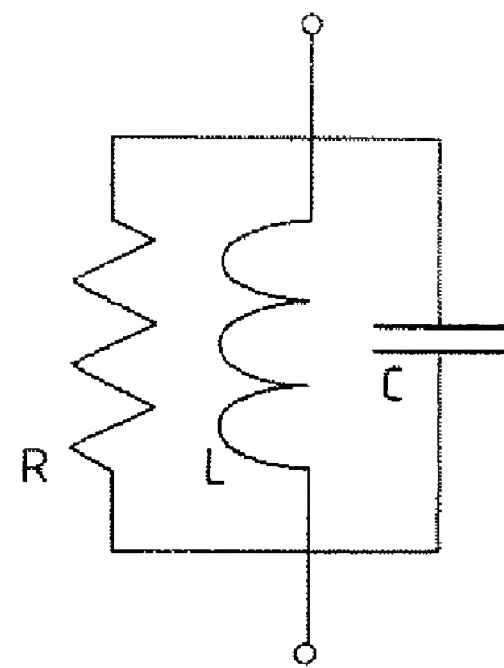


FIG. 2

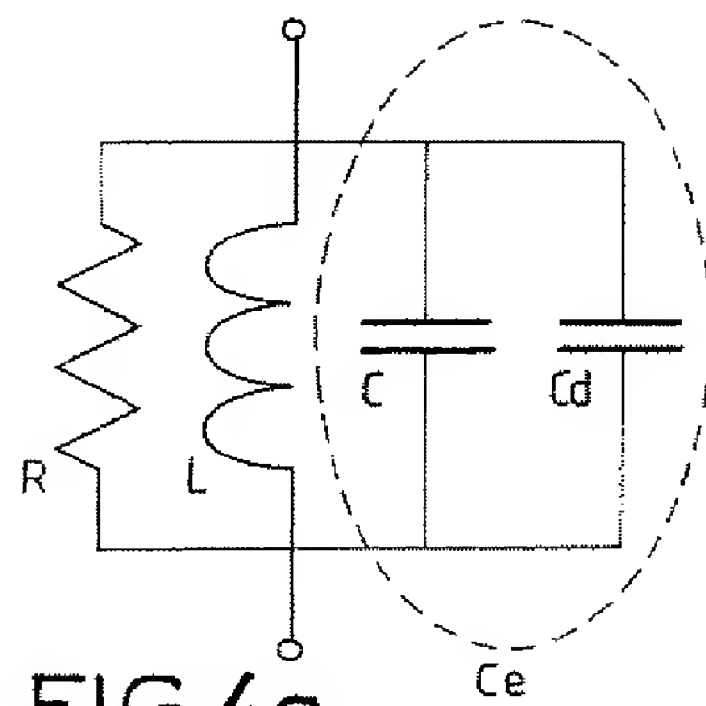


FIG. 4a

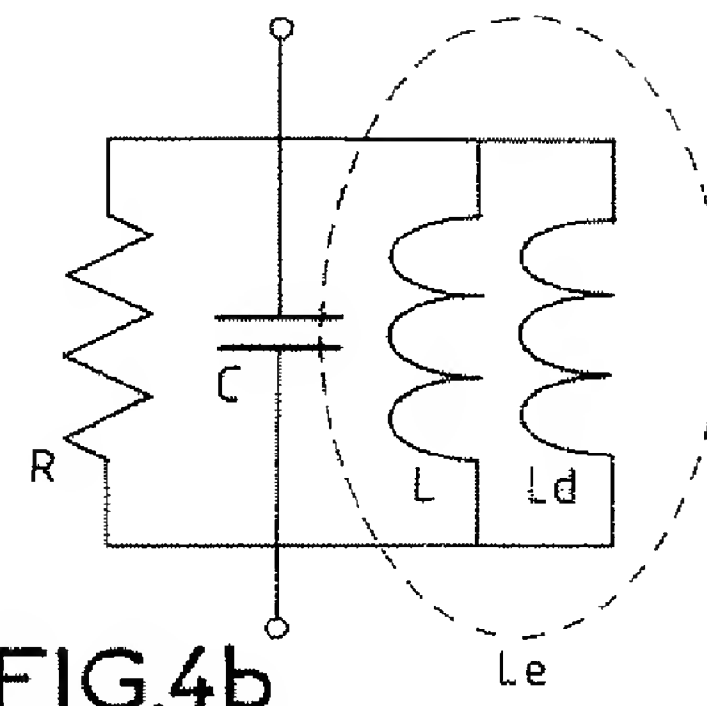


FIG. 4b

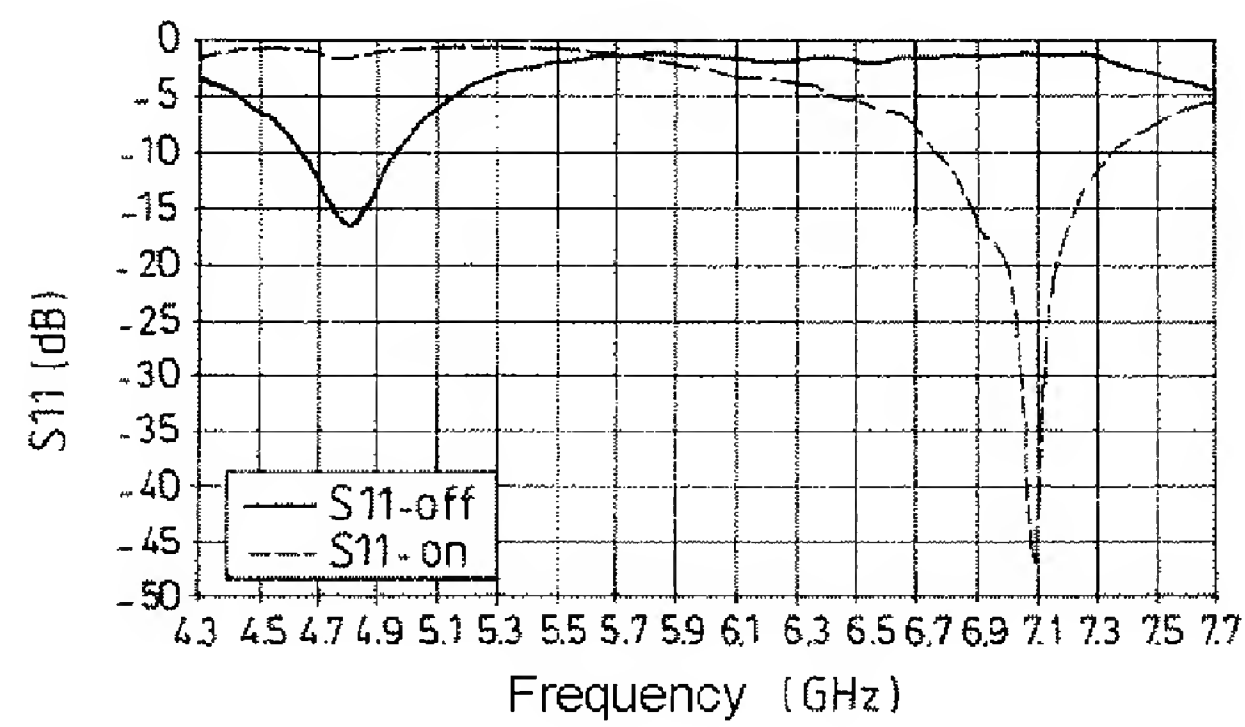


FIG. 5

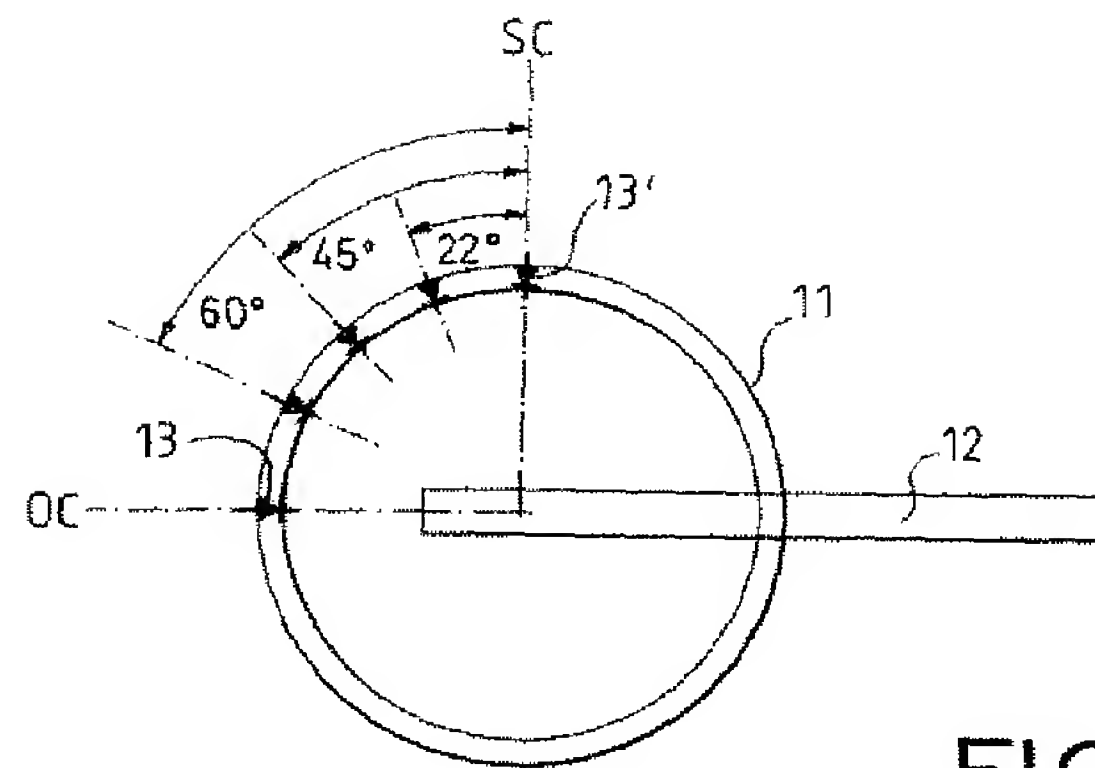


FIG. 6

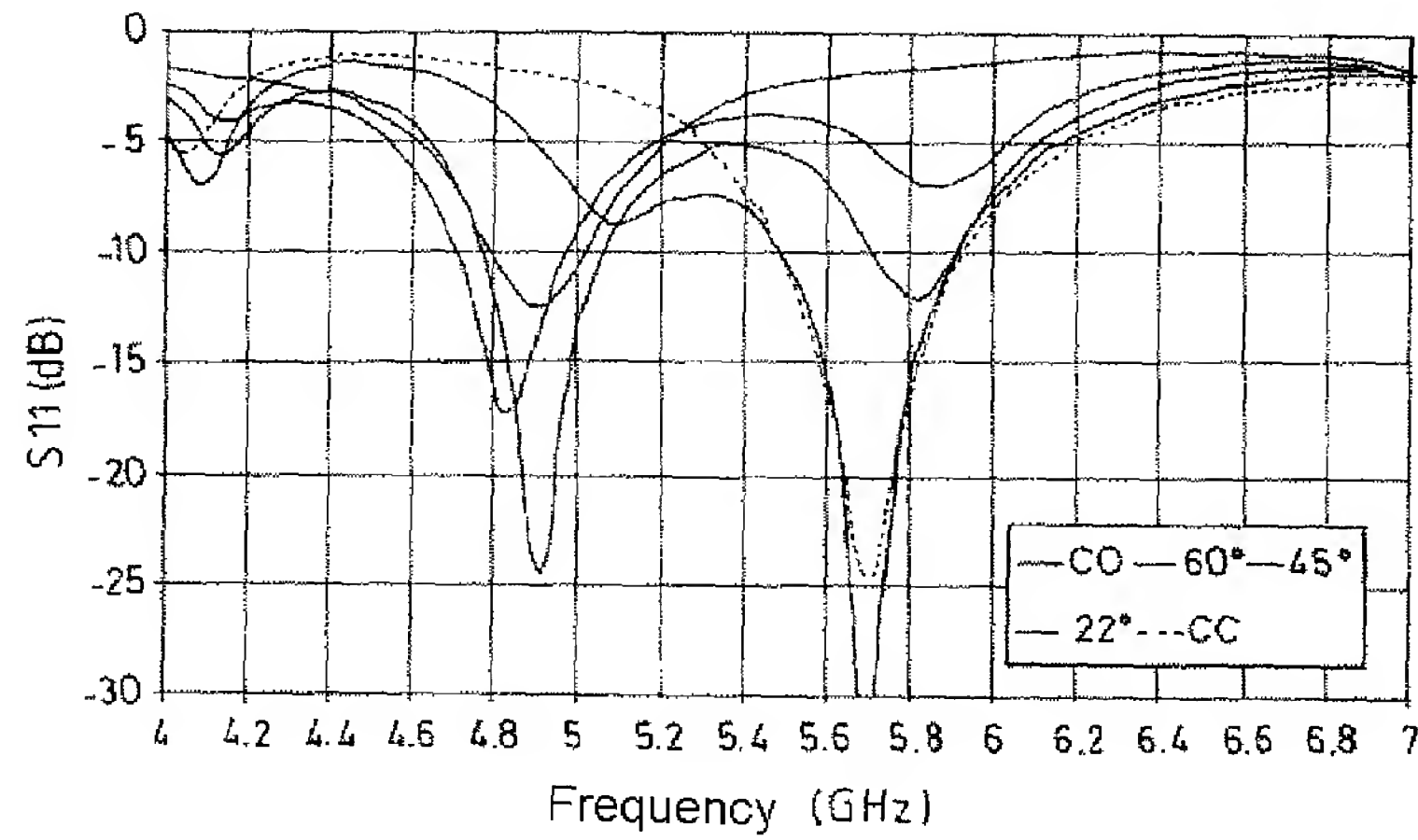


FIG. 7

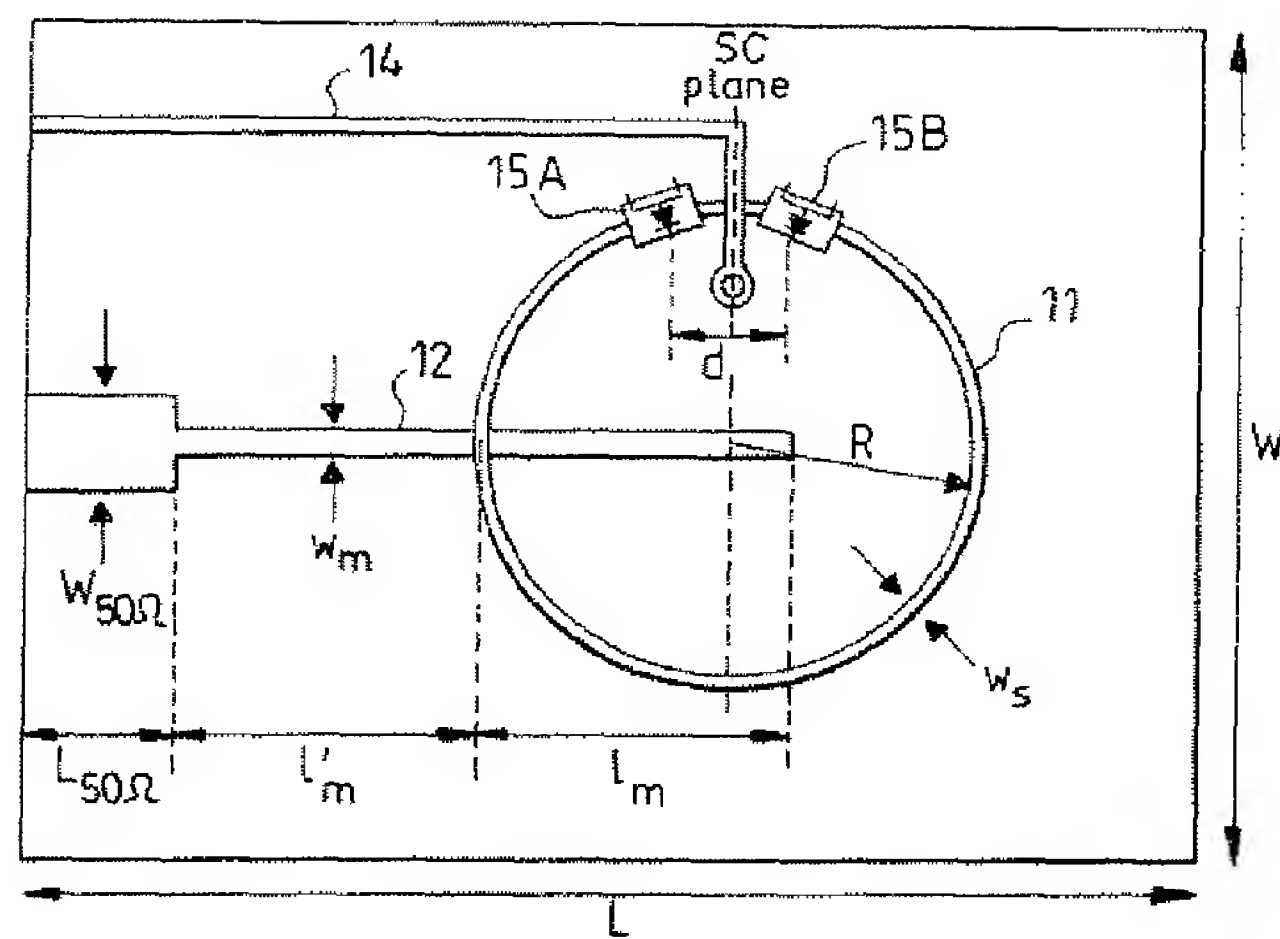


FIG. 8

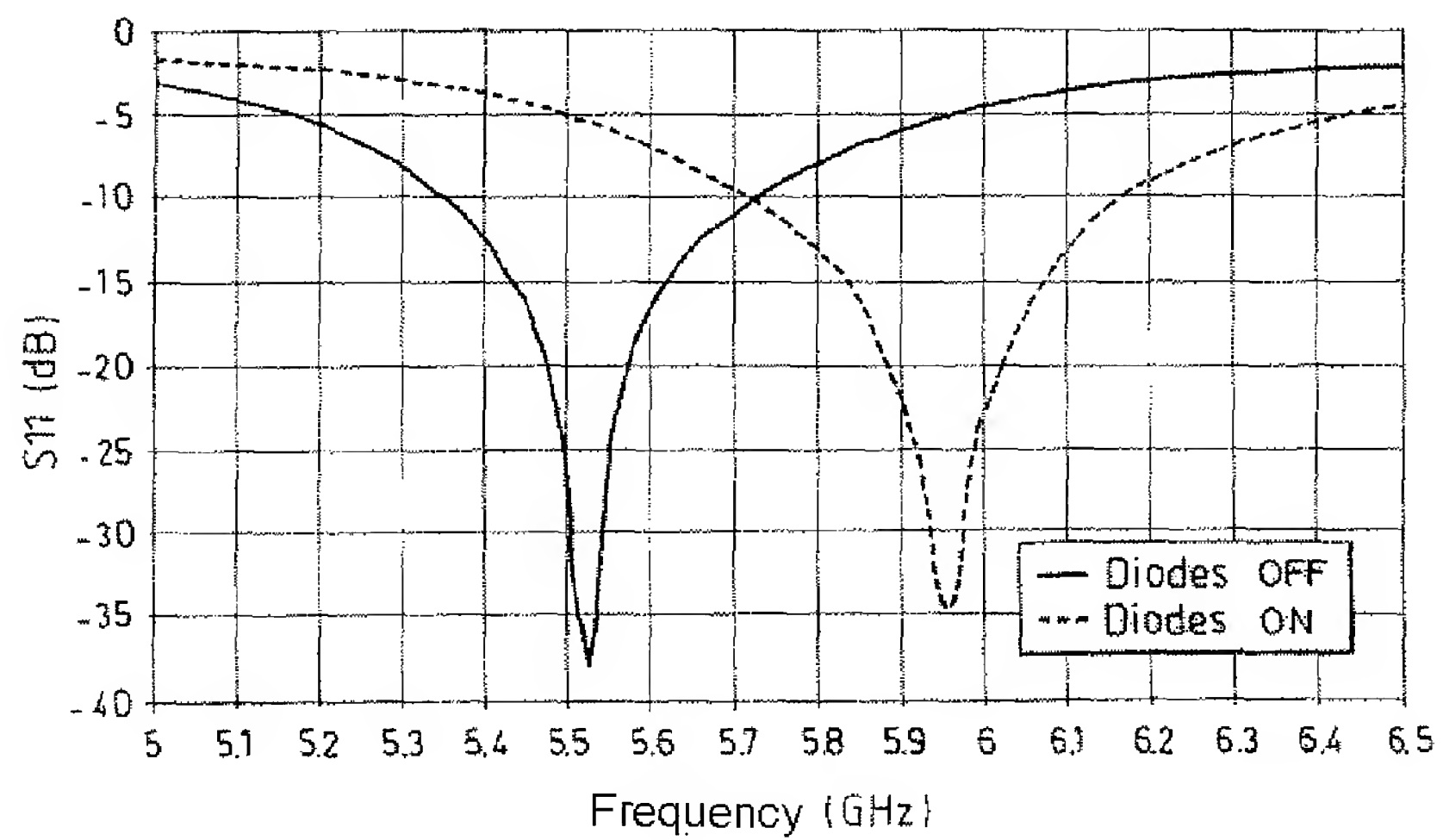


FIG. 9



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EUROPEAN SEARCH REPORT

Application Number
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DOCUMENTS CONSIDERED TO BE RELEVANT			
Category	Citation of document with indication, where appropriate, of relevant passages	Relevant to claim	CLASSIFICATION OF THE APPLICATION (Int.Cl.7)
Y	HONG CHENG-SHONG: "Small annular slot antenna with capacitor loading" ELECTRONICS LETTERS, IEE STEVENAGE, GB, vol. 36, no. 2, 20 January 2000 (2000-01-20), pages 110-111, XP006014711 ISSN: 0013-5194 * the whole document *	1-6	H01Q13/10 H01Q3/24 H01Q5/00 H01Q1/38
Y	KOLSRUD A T ET AL: "Electronically switchable slot antenna fed by microstrip line" ANTENNAS AND PROPAGATION SOCIETY INTERNATIONAL SYMPOSIUM, 1998. IEEE ATLANTA, GA, USA 21-26 JUNE 1998, NEW YORK, NY, USA, IEEE, US, 21 June 1998 (1998-06-21), pages 1180-1183, XP010292351 ISBN: 0-7803-4478-2 * the whole document *	1-6	
A	US 2002/027528 A1 (OKABE HIROSHI ET AL) 7 March 2002 (2002-03-07) * abstract; figure 9 * * column 7, paragraph 67 - paragraph 69 *	1-6	TECHNICAL FIELDS SEARCHED (Int.Cl.7) H01Q
A	US 6 028 561 A (TAKEI KEN) 22 February 2000 (2000-02-22) * abstract * * column 7, line 20 - line 67; figures 5A,B *	1-6	
A	US 5 757 326 A (FUJISAWA TERUHIKO ET AL) 26 May 1998 (1998-05-26) * column 3, line 31 - line 45 * * column 7, line 40 - line 64 *	1-6	
The present search report has been drawn up for all claims			
Place of search THE HAGUE		Date of completion of the search 18 August 2003	Examiner Angrabeit, F
CATEGORY OF CITED DOCUMENTS X : particularly relevant if taken alone Y : particularly relevant if combined with another document of the same category A : technological background O : non-written disclosure P : intermediate document T : theory or principle underlying the invention E : earlier patent document, but published on, or after the filing date D : document cited in the application L : document cited for other reasons & : member of the same patent family, corresponding document			

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Application Number
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DOCUMENTS CONSIDERED TO BE RELEVANT			
Category	Citation of document with indication, where appropriate, of relevant passages	Relevant to claim	CLASSIFICATION OF THE APPLICATION (Int.Cl.7)
A	GB 1 546 571 A (ANTENNA SPECIALISTS UK LTD) 23 May 1979 (1979-05-23) * the whole document * -----	1-6	
			TECHNICAL FIELDS SEARCHED (Int.Cl.7)
The present search report has been drawn up for all claims			
Place of search THE HAGUE		Date of completion of the search 18 August 2003	Examiner Angrabeit, F
<p>CATEGORY OF CITED DOCUMENTS</p> <p>X : particularly relevant if taken alone Y : particularly relevant if combined with another document of the same category A : technological background O : non-written disclosure P : intermediate document</p> <p>T : theory or principle underlying the invention E : earlier patent document, but published on, or after the filing date D : document cited in the application L : document cited for other reasons & : member of the same patent family, corresponding document</p>			

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18-08-2003

Patent document cited in search report	Publication date	Patent family member(s)	Publication date
US 2002027528 A1	07-03-2002	JP 2002076757 A	15-03-2002
		CN 1340981 A	20-03-2002
-----	-----	-----	-----
US 6028561 A	22-02-2000	JP 10256826 A	25-09-1998
-----	-----	-----	-----
US 5757326 A	26-05-1998	US 5940041 A	17-08-1999
		GB 2276274 A ,B	21-09-1994
		GB 2304464 A ,B	19-03-1997
		GB 2304465 A ,B	19-03-1997
		GB 2304466 A ,B	19-03-1997
		HK 1004353 A1	20-11-1998
		HK 1004354 A1	20-11-1998
		HK 1005072 A1	18-12-1998
		HK 1004355 A1	20-11-1998
		JP 7193416 A	28-07-1995
		US 5642120 A	24-06-1997
-----	-----	-----	-----
GB 1546571 A	23-05-1979	NONE	
-----	-----	-----	-----